

BJT: Fundamental relations
for DC and AC analysis
ECE352 - Spring 2007.

DC analysis

$$I_E = I_B + I_C$$

$$I_C = \alpha I_E = \beta I_B$$

$$I_E = (\beta + 1) I_B$$

$$\beta = \frac{\alpha}{1-\alpha} ; \alpha = \frac{\beta}{\beta+1}$$

AC analysis

$$V_T \approx 26 \text{ mV}$$

$$V_A \sim [50-100 \text{ V}]$$

↑
Early voltage

$$r_{\pi} = \frac{V_T}{I_B}$$

$$r_e = \frac{V_T}{I_E}$$

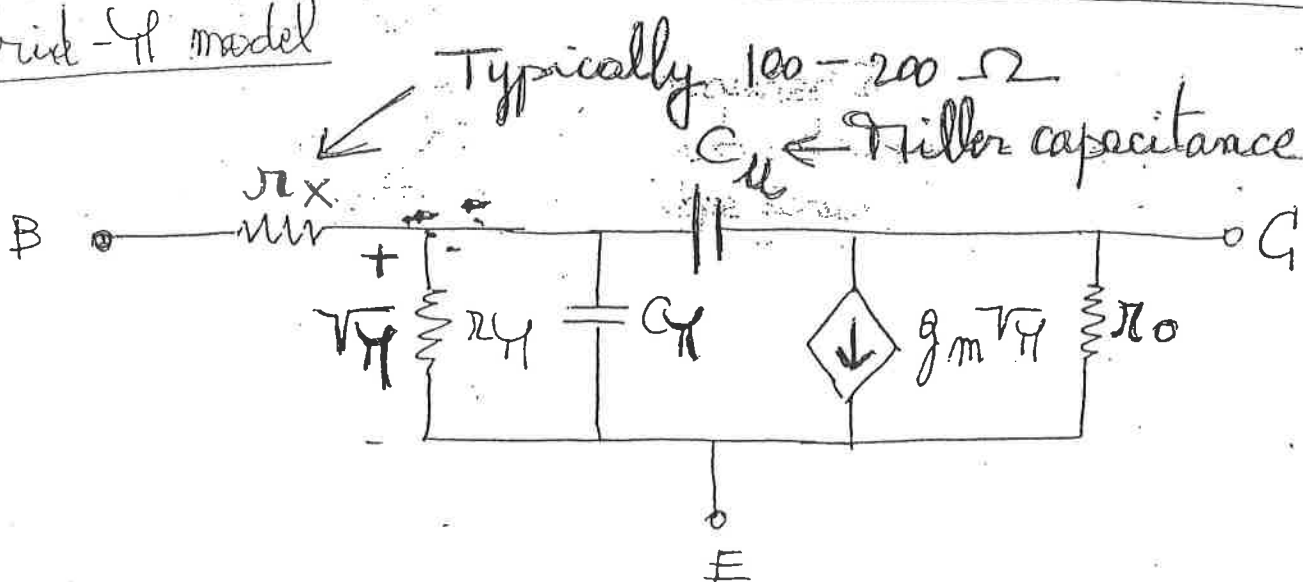
$$r_{\pi} = (\beta + 1) r_e$$

$$g_m = I_C / V_T$$

$$r_{\pi} g_m = \beta ; r_e g_m = \alpha$$

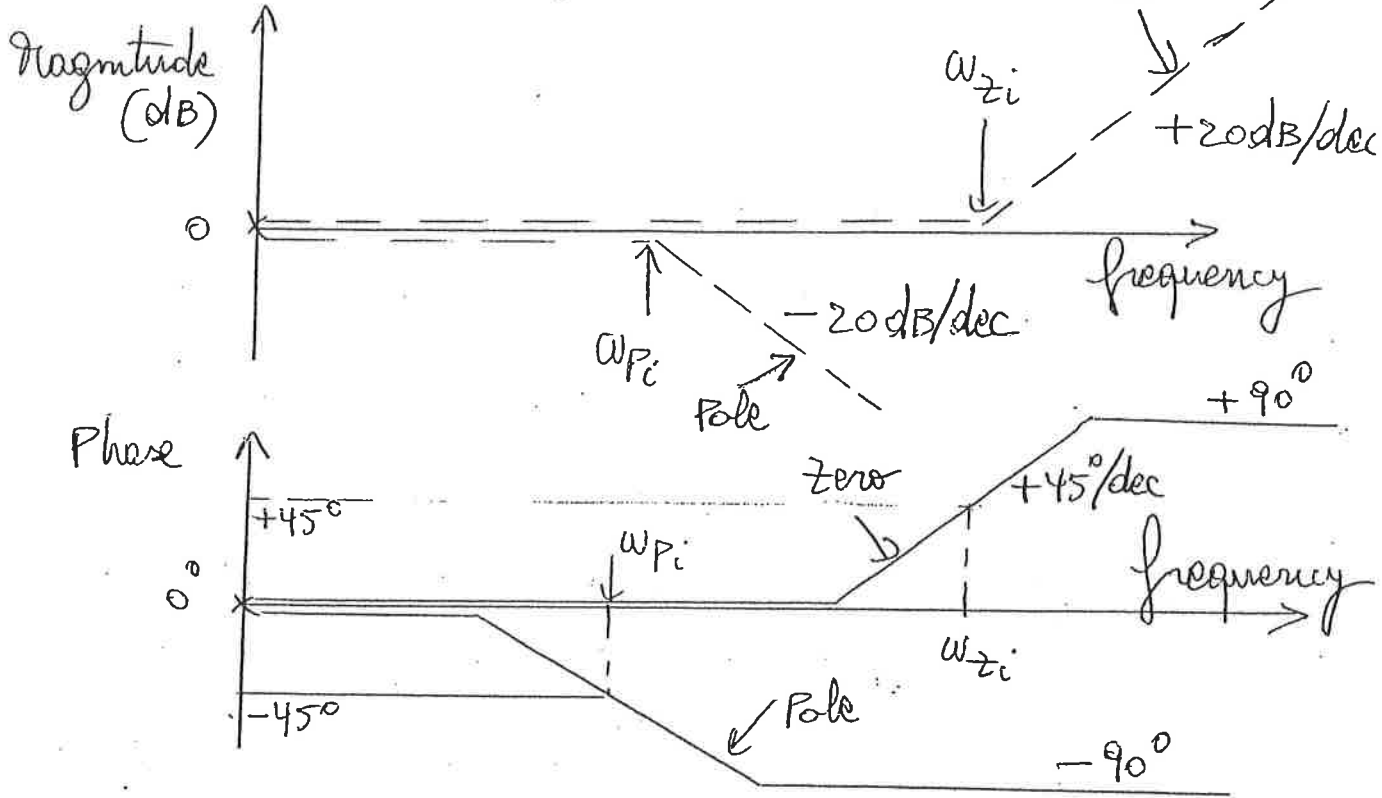
$$r_o = V_A / I_C$$

Hybrid- π model

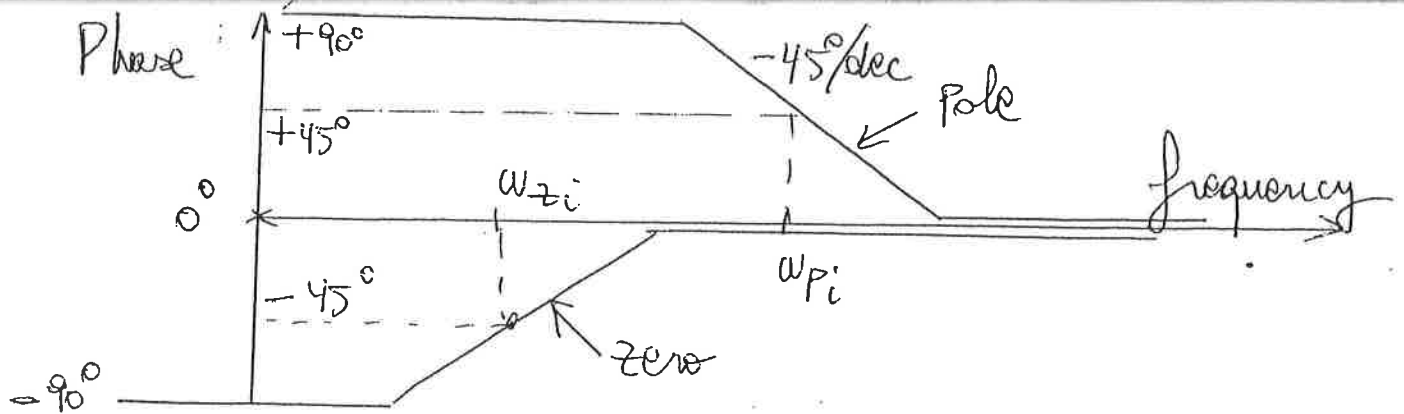
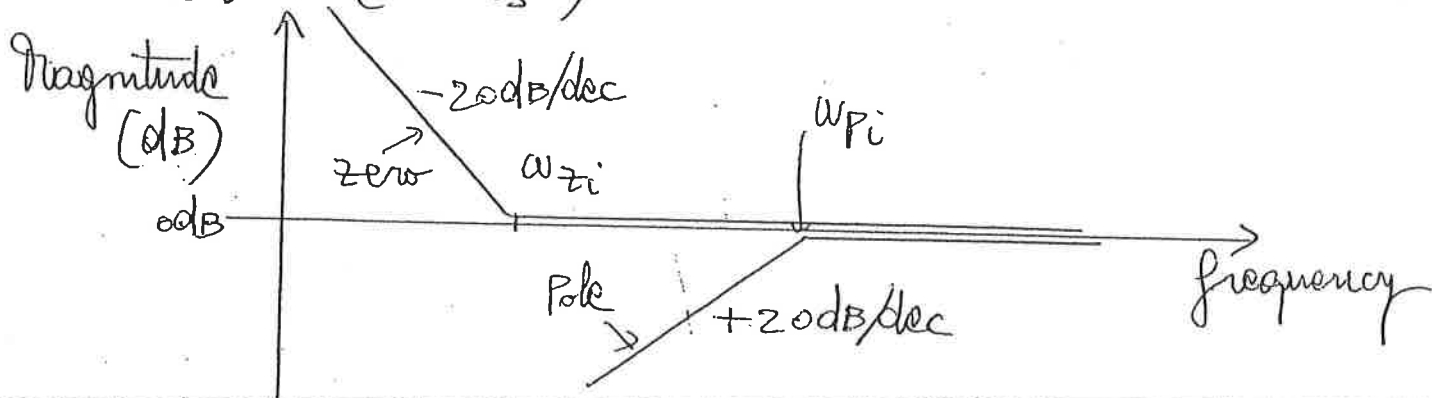


Summary Sheet

$$G(s) = \frac{(1 + \frac{s}{\omega_{zi}}) \dots (1 + \frac{s}{\omega_{zm}})}{(1 + \frac{s}{\omega_{pi}}) \dots (1 + \frac{s}{\omega_{pm}})} \quad \begin{array}{l} n \text{ poles} \\ m \text{ zeros} \end{array}$$



$$F_L(s) = \frac{(1 + \frac{s}{\omega_{zi}})}{(1 + \frac{s}{\omega_{pi}})} \quad \begin{array}{l} n \text{ poles} \& \ m \text{ zeros} \end{array}$$



Basic Types of Feedback Amplifiers

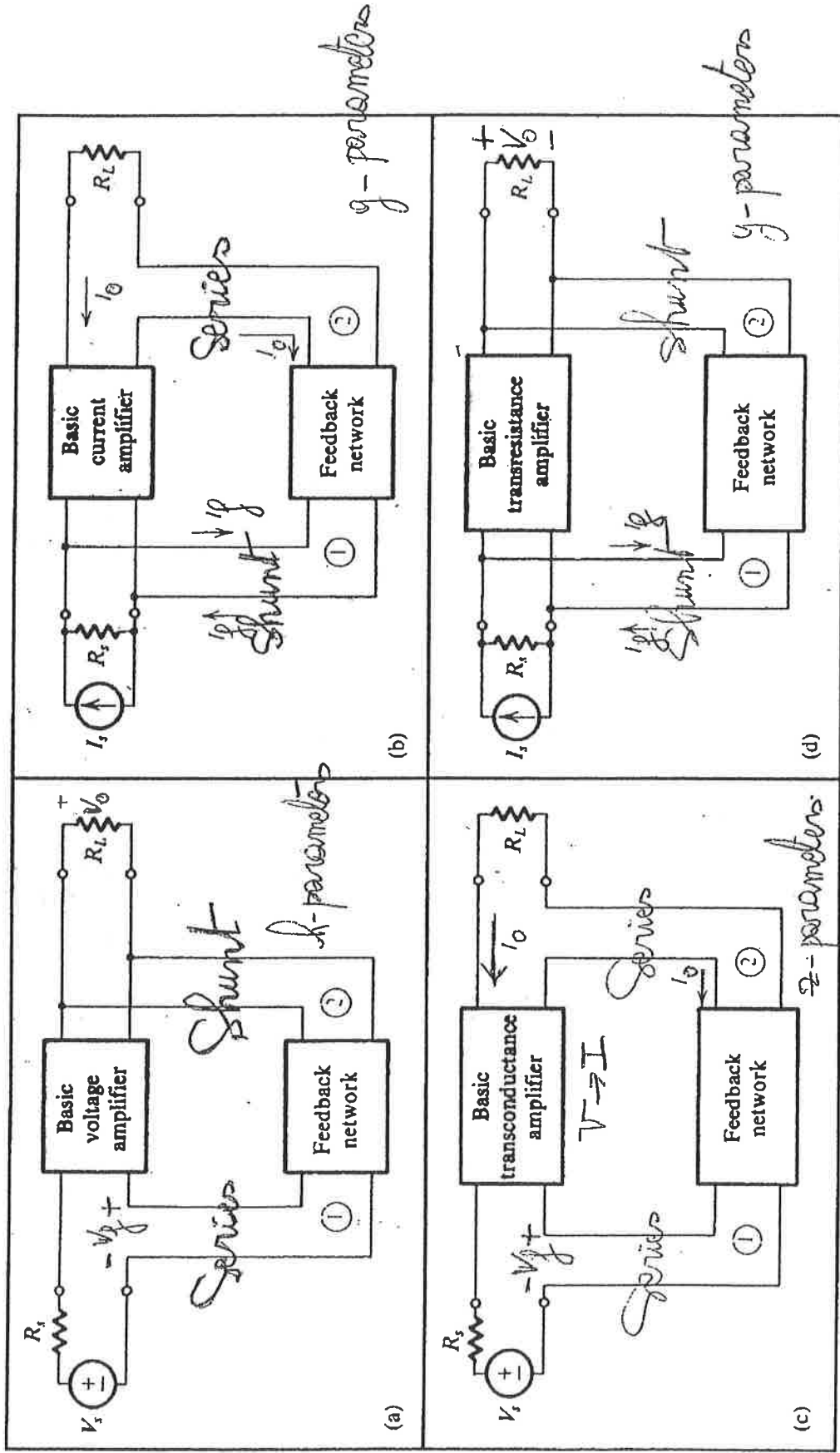
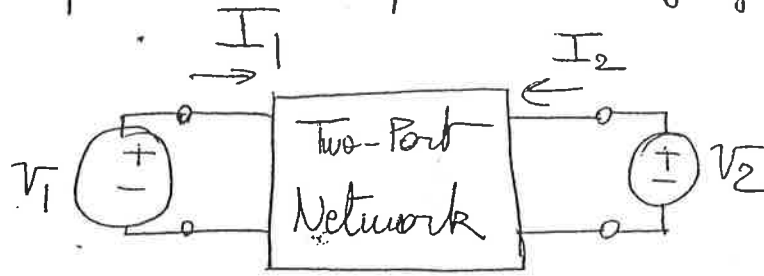


Fig. 8.4 The four basic feedback topologies: (a) voltage-sampling series-mixing (series-shunt) topology; (b) current-sampling shunt-mixing (shunt-series) topology; (c) current-sampling series-mixing (series-series) topology; (d) voltage-sampling shunt-mixing (shunt-shunt) topology.

Equivalent Two-port Network for feedback circuits.



APPENDIX B TWO-PORT NETWORK PARAMETERS

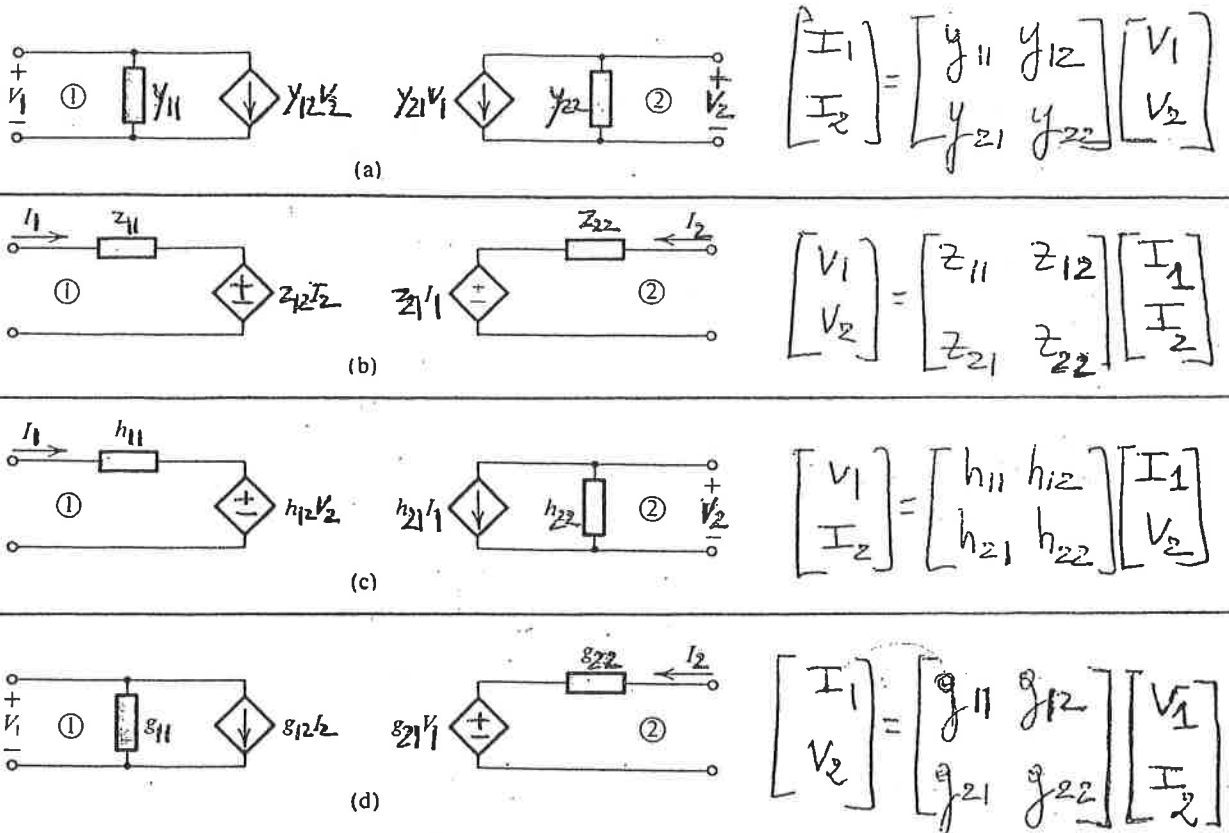


FIGURE B.6 Equivalent circuits for two-port networks in terms of (a) y , (b) z , (c) h , and (d) g parameters.

First Order Filter Functions

* First order filter functions are of the form

$$T(s) = \frac{a_1 s + a_0}{s + \omega_0} = \frac{a_1 \left(s + \frac{a_0}{a_1} \right)}{s + \omega_0}$$

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Low Pass

Filter Type and $T(s)$	s -Plane Singularities	Bode Plot for $ T $	Passive Realization	Op Amp-RC Realization
(a) Low-Pass (LP) $T(s) = \frac{a_0}{s + \omega_0}$			<p> $CR \approx \frac{1}{\omega_0}$ dc gain ≈ 1 </p>	<p> $CR_2 \approx \frac{1}{\omega_0}$ dc gain $\approx -\frac{R_2}{R_1}$ </p>
(b) High-Pass (HP) $T(s) = \frac{a_1 s}{s + \omega_0}$			<p> $CR \approx \frac{1}{\omega_0}$ High-frequency gain = 1 </p>	<p> $CR_2 \approx \frac{1}{\omega_0}$ High-frequency gain $\approx -\frac{R_2}{R_1}$ </p>

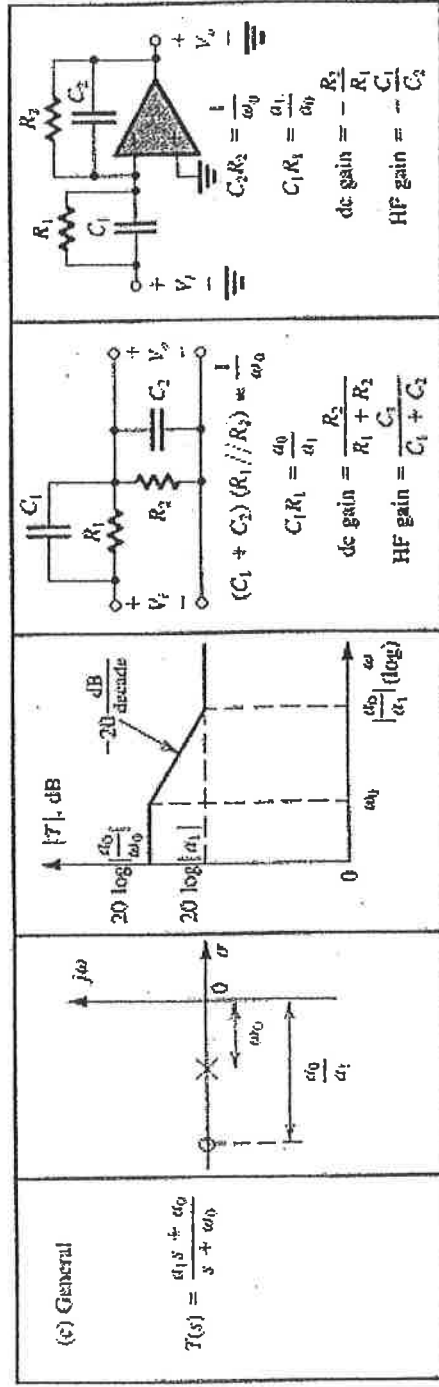
High Pass

First Order Filter Functions

* First order filter functions are of the form

$$T(s) = \frac{a_1 s + a_0}{s + \omega_0}$$

General



All Pass

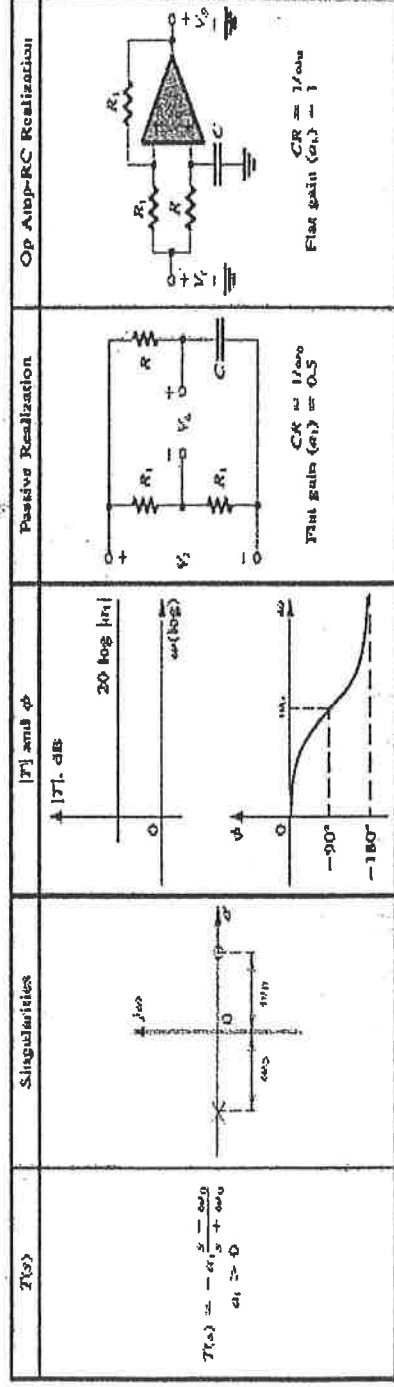


Fig. 11.14 First-order all-pass filter.

Passive Second Order Filter Functions

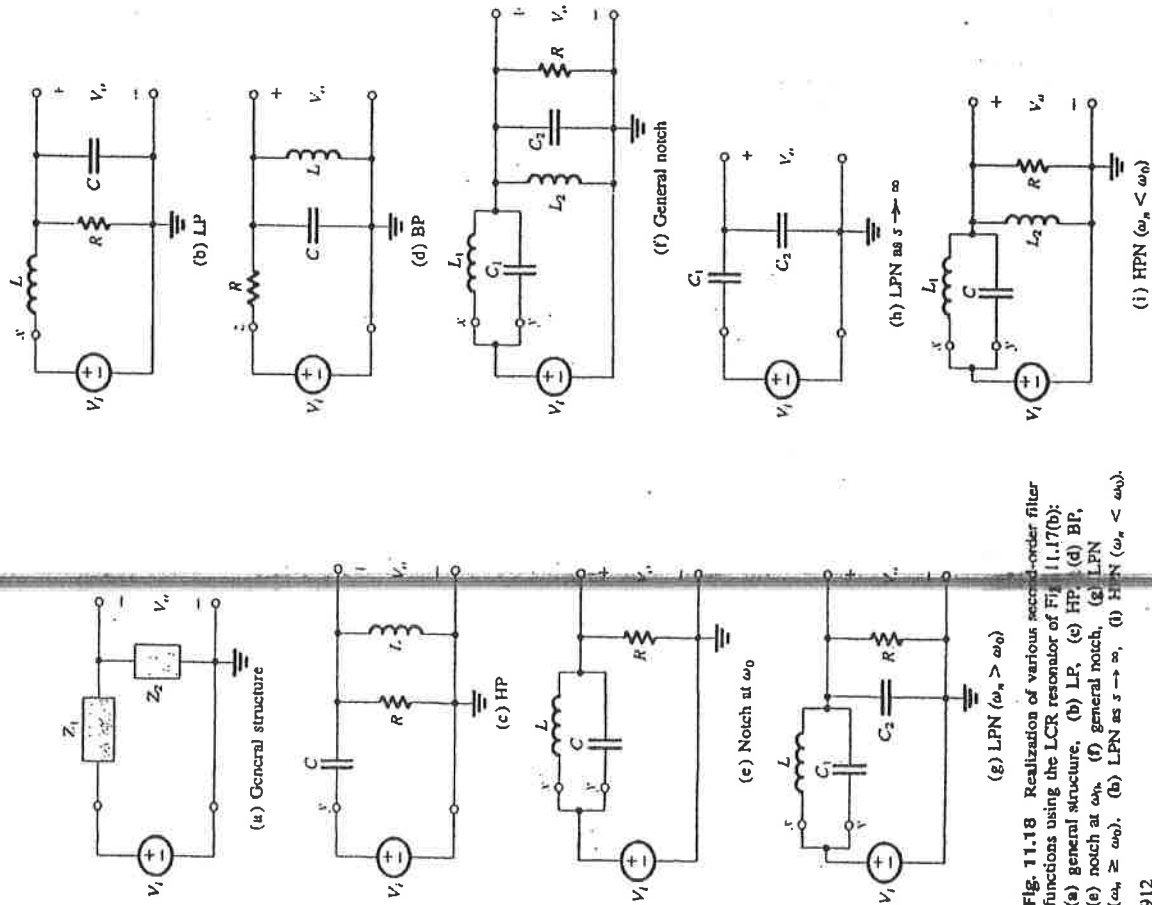


Fig. 11.18 Realization of various second-order filter functions using the LCR resonator of Fig. 11.17(b): (a) general structure, (b) LP, (c) HP, (d) BP, (e) notch at ω_0 , (f) general notch, (g) LPN ($\omega_n \approx \omega_0$), (h) LPN as $s \rightarrow \infty$, (i) HPN ($\omega_n < \omega_0$).

* Second order filter functions can be implemented with simple RLC circuits

* General form is that of a voltage divider with a transfer function given by

$$T(s) = \frac{V_o(s)}{V_i(s)} = \frac{a_2s^2 + a_1s + a_0}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2}$$

* Seven types of second order filters

- High pass
- Low pass
- Bandpass
- Notch at
- General notch
- Low pass notch (2 types)
- High pass notch

Second-Order Filter Functions

$$T(s) = \frac{a_2 s^2 + a_1 s + a_0}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

Notch

$$a_1 = 0$$

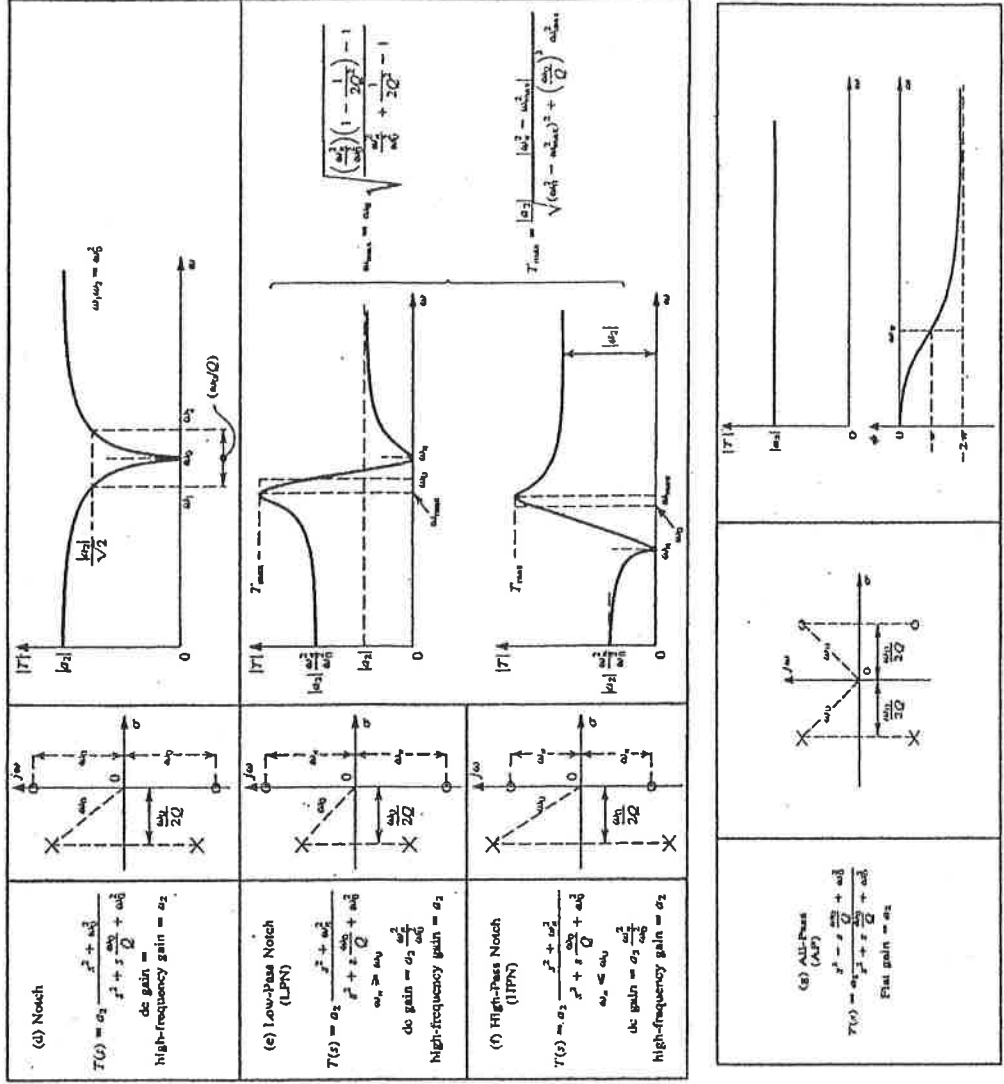
Low Pass Notch

$$a_1 = 0$$

High Pass Notch

$$a_1 = 0$$

All-Pass



Second-Order Filter Functions

$$T(s) = \frac{a_2 s^2 + a_1 s + a_0}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

Low Pass

$a_1=0, a_2=0$

High Pass

$a_0=0, a_1=0$

Bandpass

$a_0=0, a_2=0$

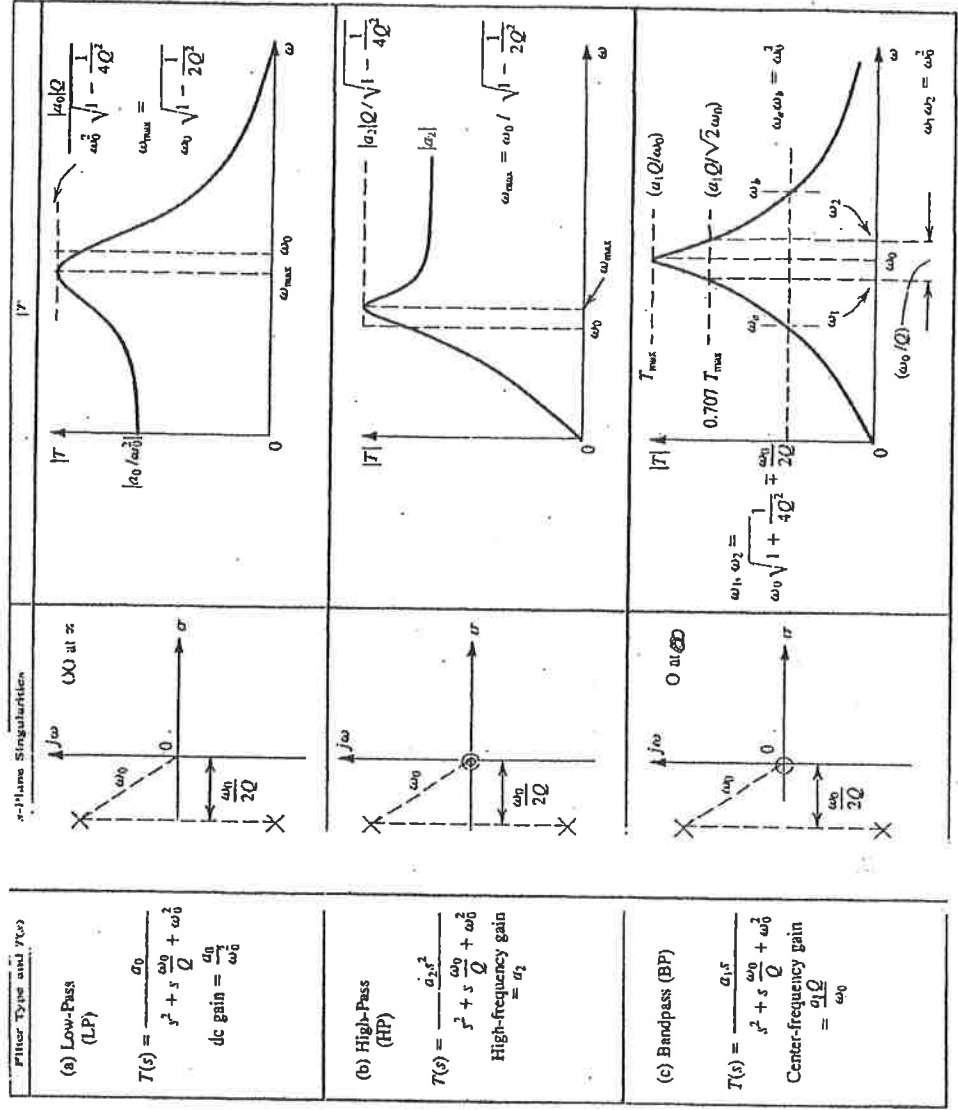


Fig. 11.16 Second-order filtering functions.