

Detectability

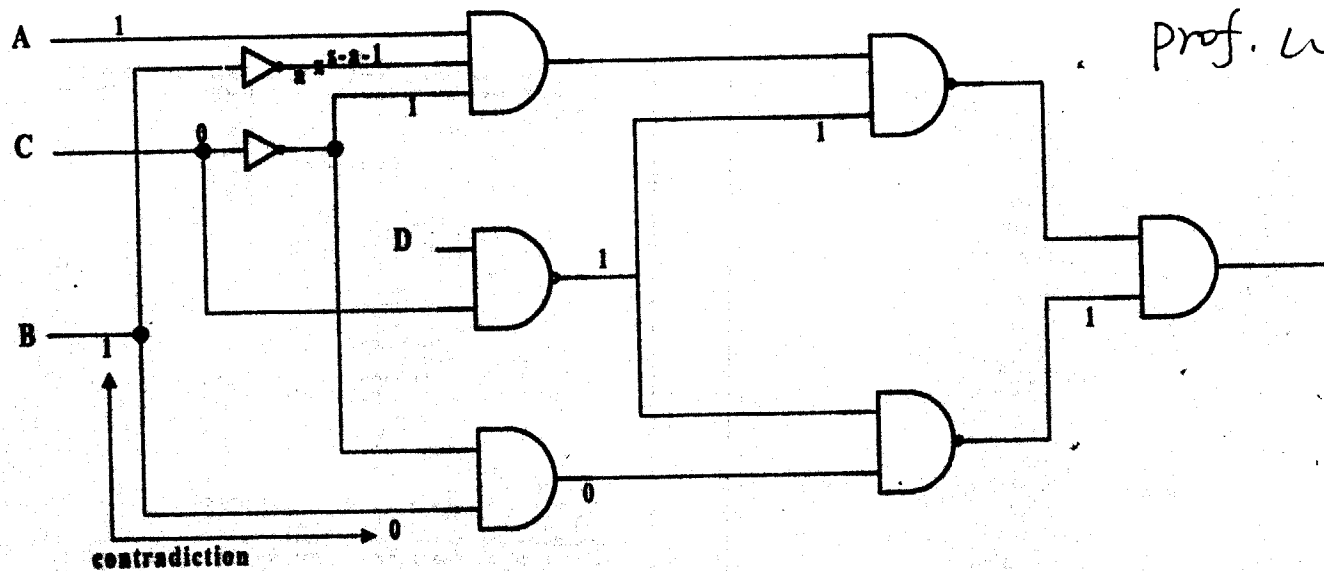
A fault f is detectable if there exists a test t which detects f .

- Example :
a s-a-1 in (a) is not detectable! Why?

ECES 682

VLSI Test & Validation

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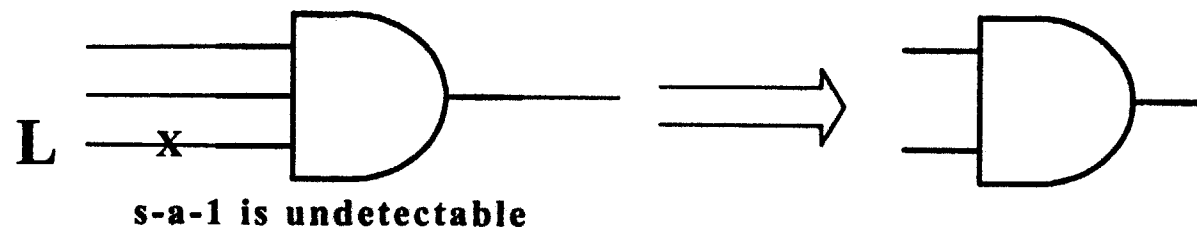


Redundancy

A combinational circuit containing an undetectable stuck-at fault is called redundant.

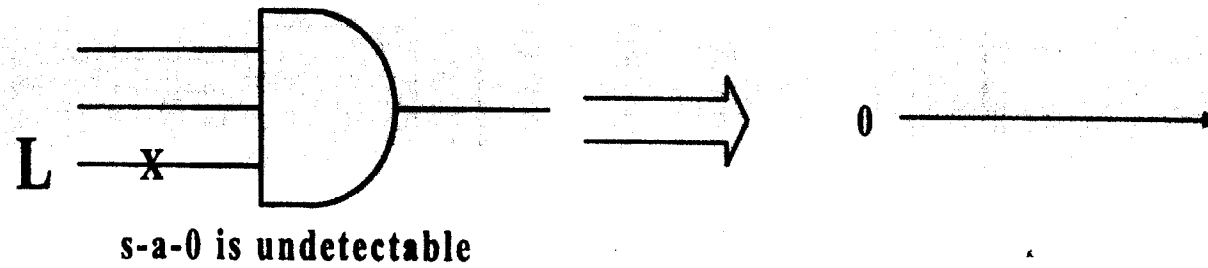
Reason: there are some unnecessary lines or gates.

- Example :



Since line L can be always 1 without changing the function, line L can be removed.

Another Example



Some other properties of Redundancy

- ① If f is detectable, g is undetectable
 $\Rightarrow f$ may become undetectable in the presence
of g
 $\therefore f$ is called a 2nd generation redundant fault
- ② Two undetectable single faults f and g may
become detectable, if they are present
simultaneously.

Important thing

How to recognize redundant fault , s.t. we don't need to try all input combinations for a redundant fault?

★ No difference between an undetectable fault and hard-to-detect fault.

try long cpu time still can't find a test pattern

★ $\text{complexity}(\text{test generation}) = \text{complexity}(\text{identify redundant fault}) = \text{NP-complete.}$

★ Although test generation problem is NP-complete, practical test generation algorithms run in polynomial time.

Say n^3 get coverage 99% for very large circuits.

★ NP-complete means no polynomial time algorithm can find test patterns for all circuits.

★ Redundant fault is usually the major factor for the worst case.

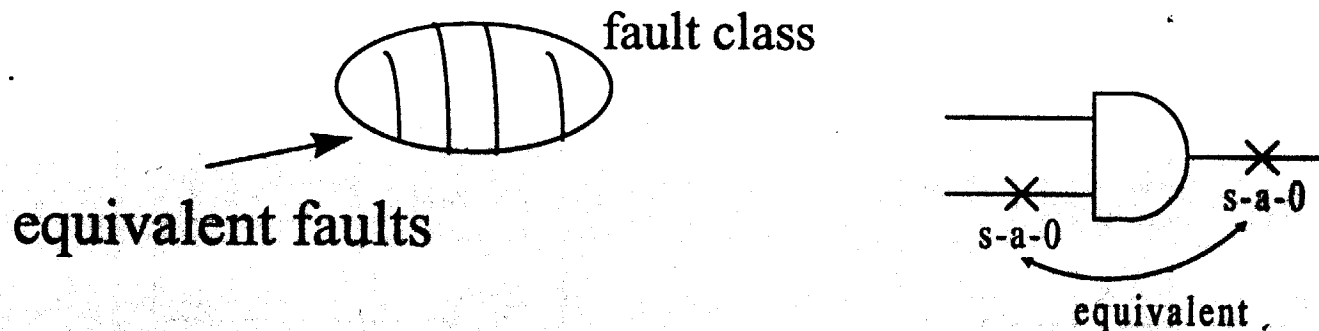
Fault equivalence and fault Location

combinational ckts

Def : Two faults f and g are functional equivalent $\leftrightarrow Z_f(x) = Z_g(x)$ for all x

Def : A test t is said to distinguish two faults f and g if $Z_f(t) \neq Z_g(t)$

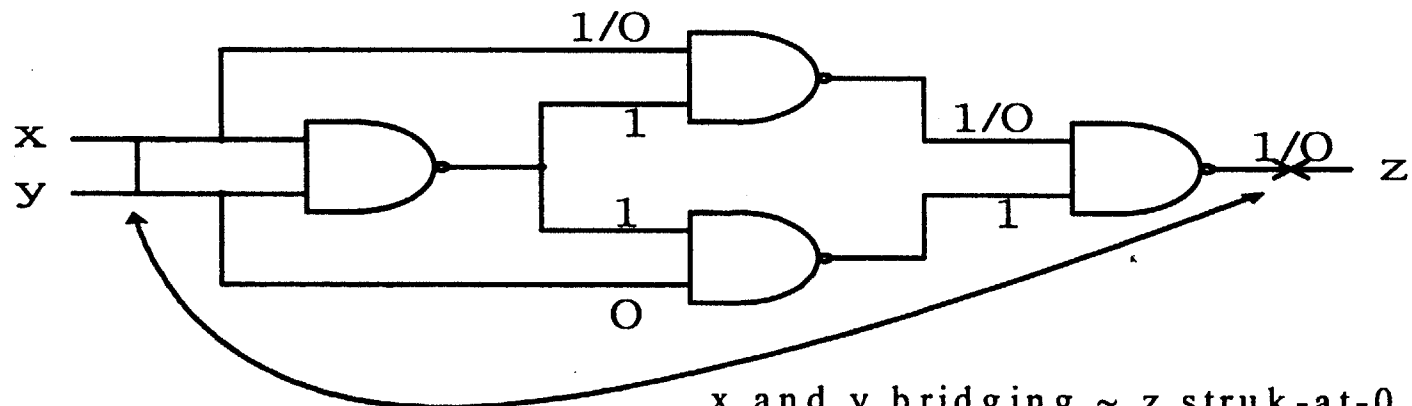
Example :



★ just need to detect each class , instead of each fault

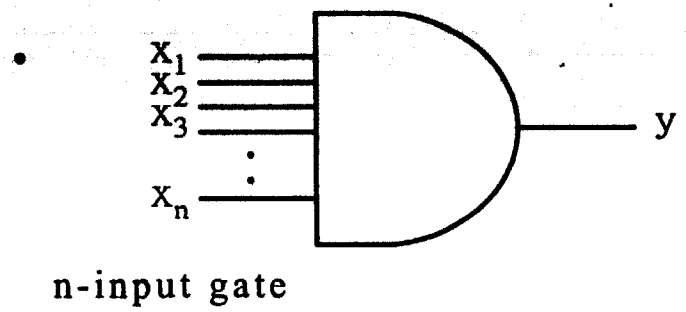
Equivalence relation is not restricted in the same fault type

example :



x and y bridging \approx z stuck-at-0.
(Wired-AND)

Example : $x=1$, $y=0$, $z=0$, but z should be a 1



We have $2(n+1)$ signal stuck-at faults, but just need to consider only $n+2$ signal stuck-at faults.

n-input gate

$X_1 \dots X_n$, s-a-0 \approx y s-a-0 \therefore one is enough (y s-a-0) (1) fault

X_1 : stuck-at-1	}	<u>(n+1)</u> faults
X_2 : .		
X_3 : .		
.		
X_n : stuck-at-1		
Y .		

\therefore n+2 faults

- The process of reducing faults by equivalence function is called **fault collapsing**
- fault location : discussed latter

Def : Two faults f and g are functionally equivalent

$$\leftrightarrow R_f(q_{1f}, T') = R_g(q_{1g}, T') \text{ for any } T'$$

Note : If faults prevent initialization

\Rightarrow this relation does not hold.

Def : f and g are functionally equivalent under test sequence

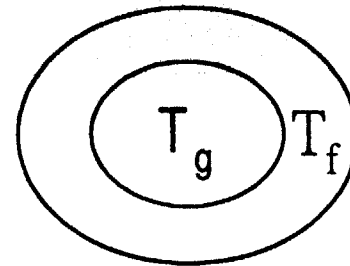
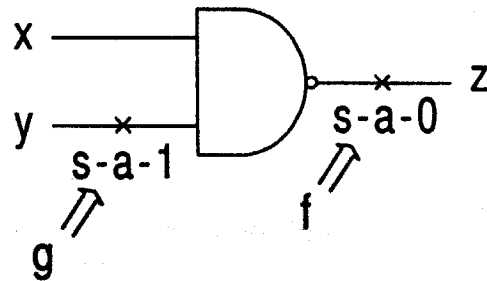
$$T = \{T_1, T'\} \leftrightarrow R_f\{q_{1f}, T'\} = R_g\{q_{1g}, T'\}$$

Fault Dominance

Def : Let T_g be the set of tests that detect a fault g . We say that a fault f dominates the fault $g \leftrightarrow f$ and g are functionally equivalent under T_g .

\therefore If f dominates $g \Rightarrow$ any set detects g also detects f .

Example :



$$T_g = \{ 10 \}$$

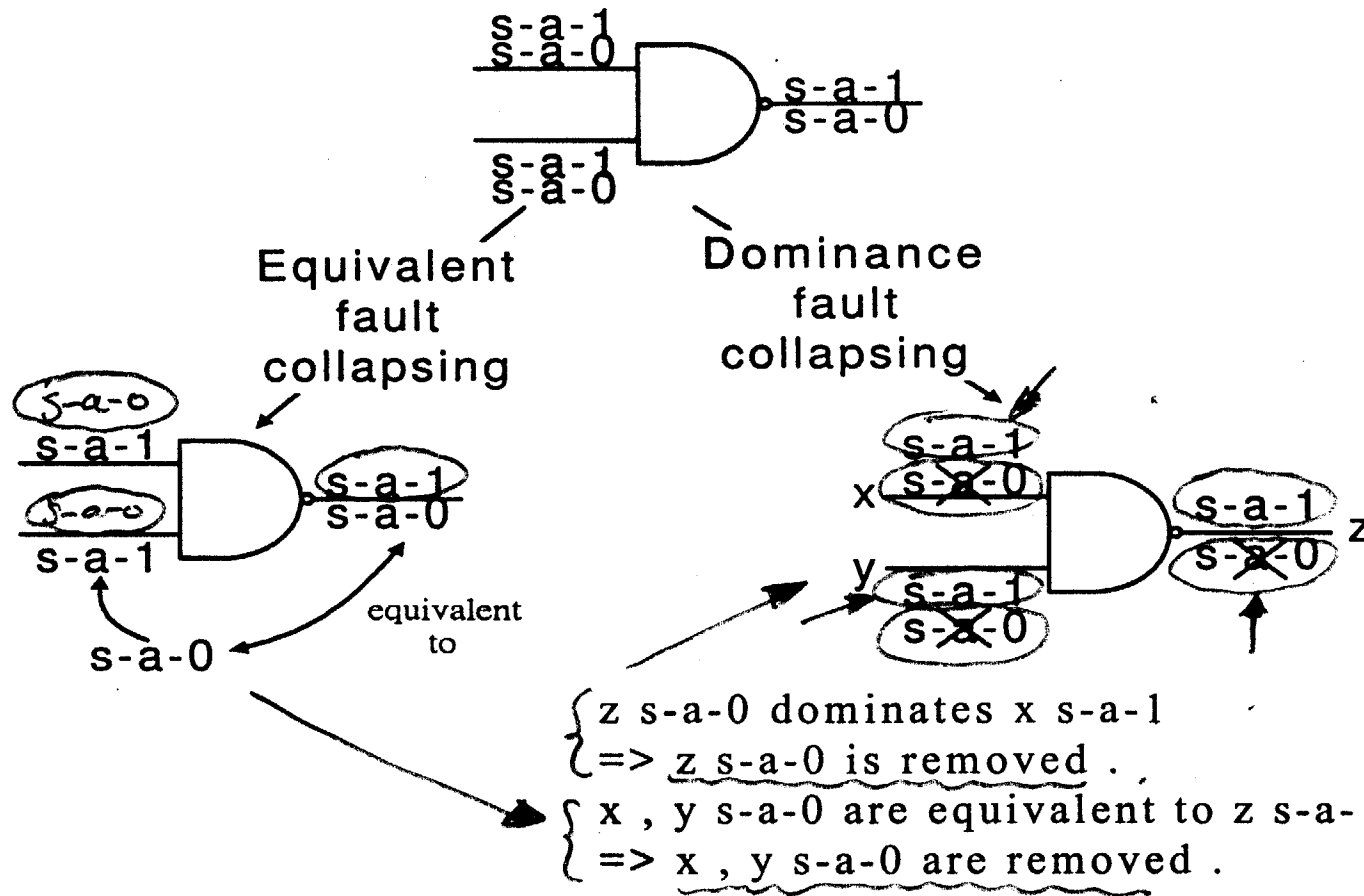
$$T_f = \{ 10, 01, 00 \}$$

$\therefore f$ dominates g .

$\therefore f$ s-a-0 can be removed, since a test set detects y s-a-1 definitely detects z , s-a-0.

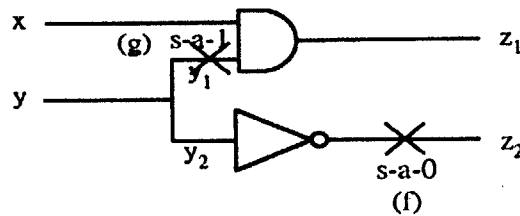
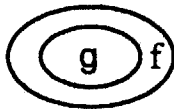
This is called **dominance fault collapsing**.

Example :



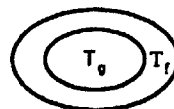
Note : We have two faults f and g such that any test detects g also detects f , without f dominating g .

example:



$T_g = \{ 1 0 \}$ also detects f

$T_f = \{ x 0 \}$



but under $xy = 10$, f and g are not functionally equivalent.

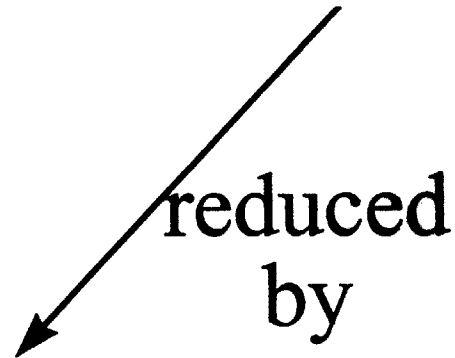
Note : f dominant $g \rightarrow$ under T_g f and g are equivalent \implies

~~does not necessarily hold !!~~

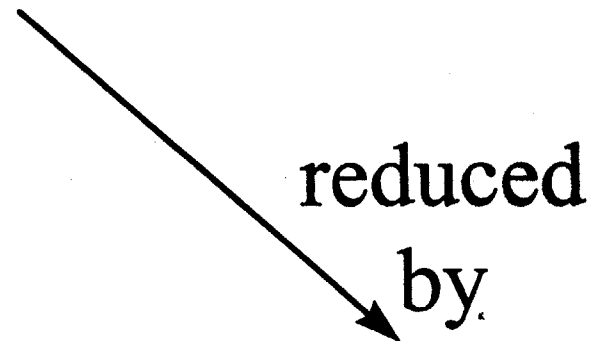
The Single Stuck-at Fault Model

- Most widely used model
- Represents many physical faults

Fault list



Fault equivalence
relation



Fault dominance
relation

Fault Equivalence relation

Equivalent fault

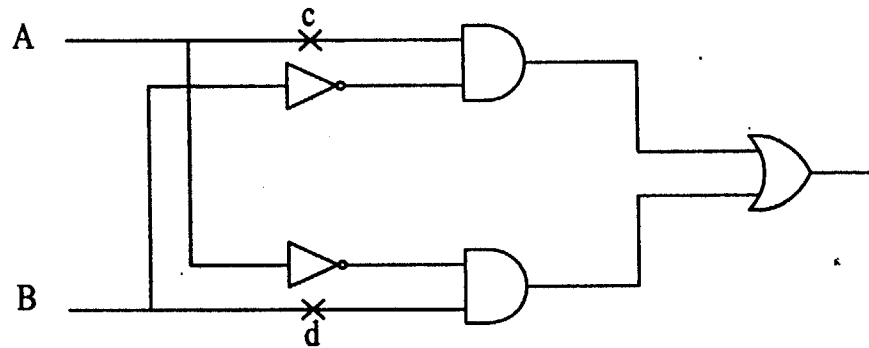


Can be discussed from
Functional Equivalence
Relation

Or from Structure
Equivalence Relation

Functional equivalent relation detection is NP complete

Example :

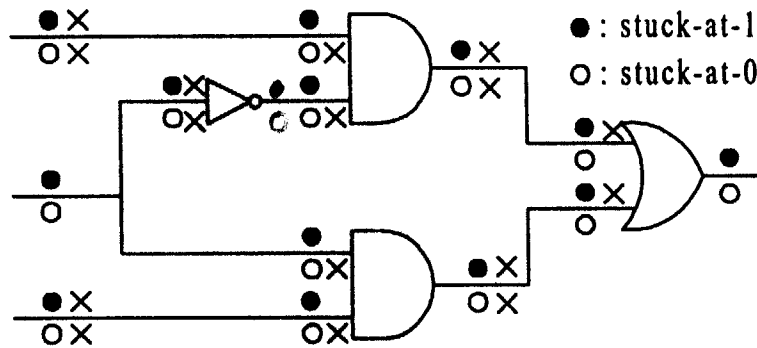


- c s-a-1 and d s-a-1 are functionally equivalent
- There is no structural relation between c and d ,
So no simple way to determine the equivalence
of c and d s-a-1

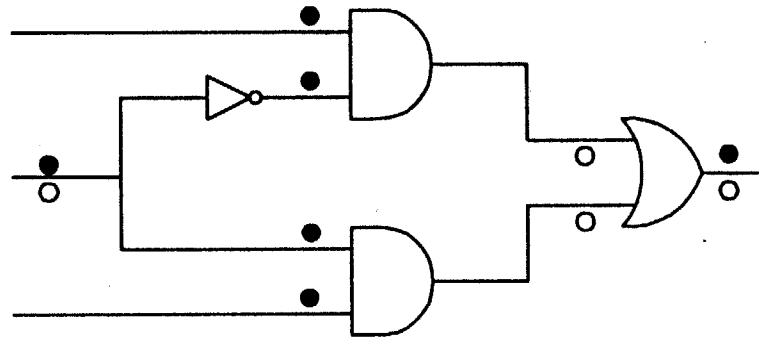
- Structure equivalence analysis: local analysis based on structure of the circuit
- Functional equivalence analysis: global analysis (entire function)

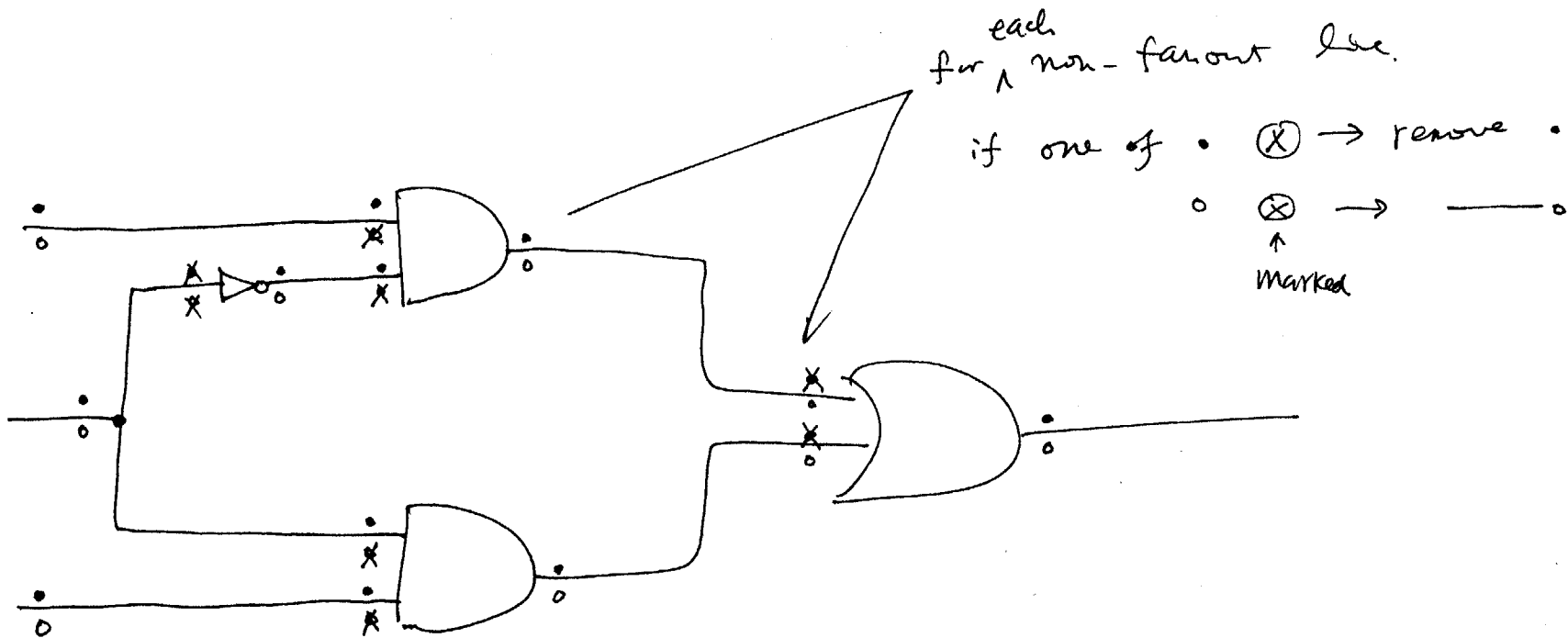
Structure Analysis

example :



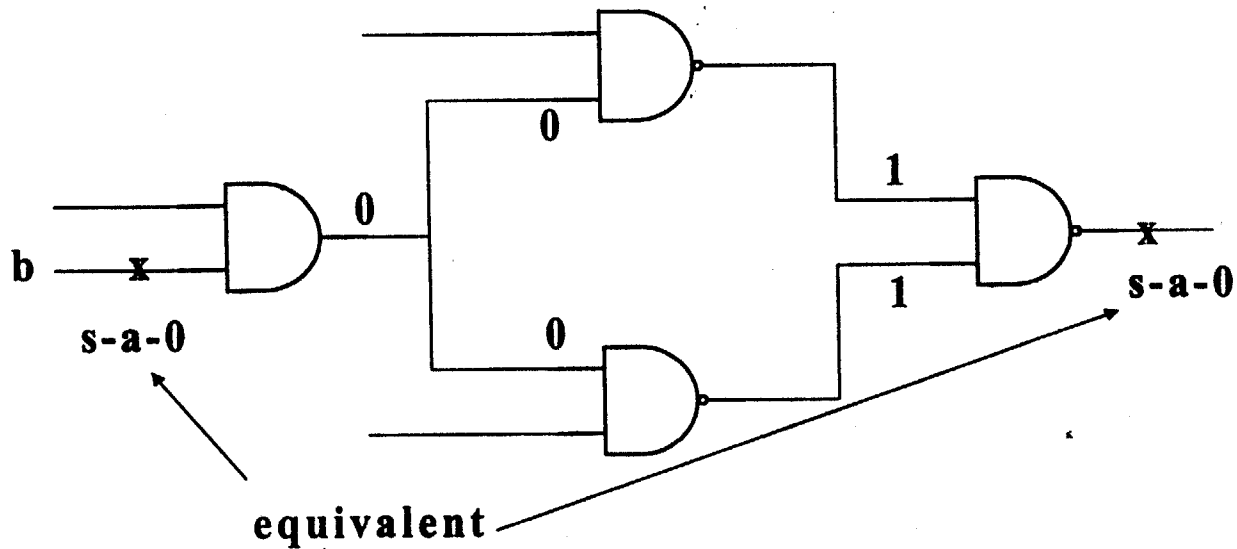
⇒



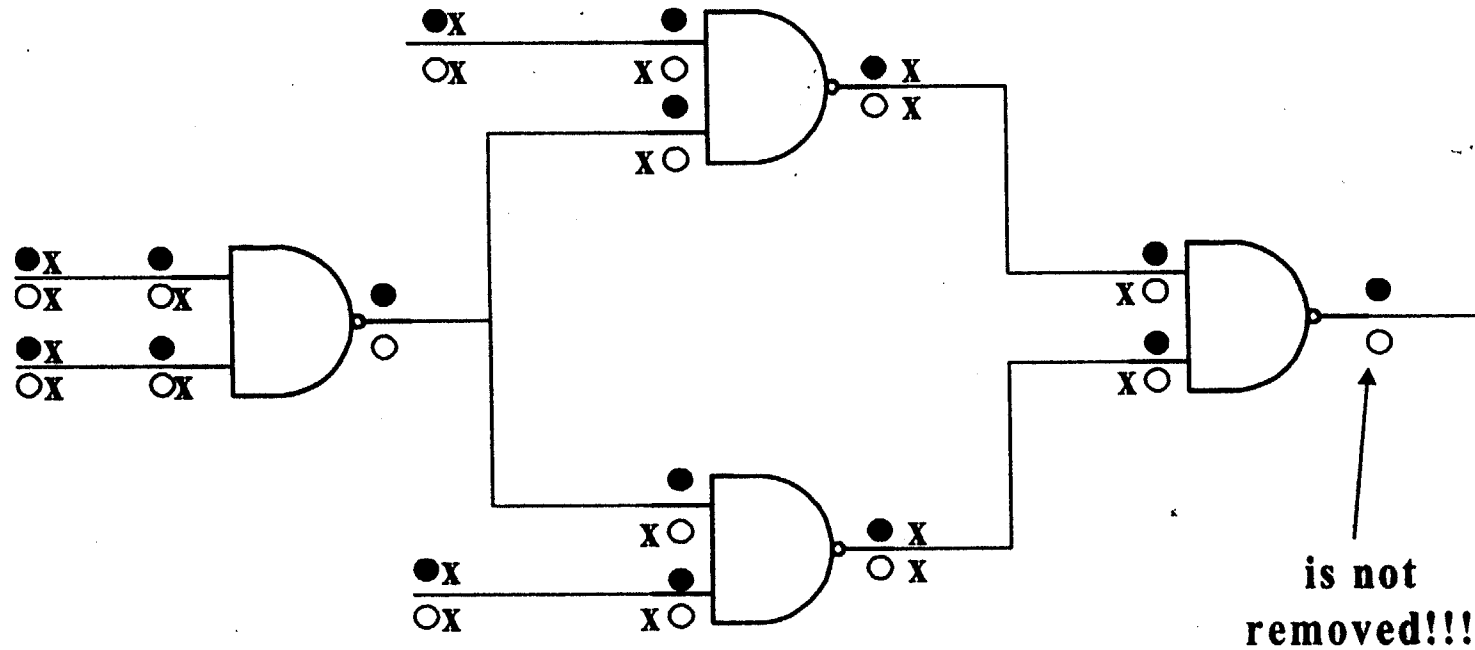


use relation of purely equivalent relation

- Example :



But, our labeling technique can not find this, since reconvergent fanout is not taken into account.



* The process does not keep track of the fanout.
reconvergent information

- The obtained structural equivalence class is not maximal
- Try to expand the power of labeling technique?
NO! The gain achievable does not justify the cost of additional cost.
(i.e., spend so much effort, might just find very limited # of equivalent faults.)
- The structural fault collapsing reduces the initial fault list by 50% (on the average).

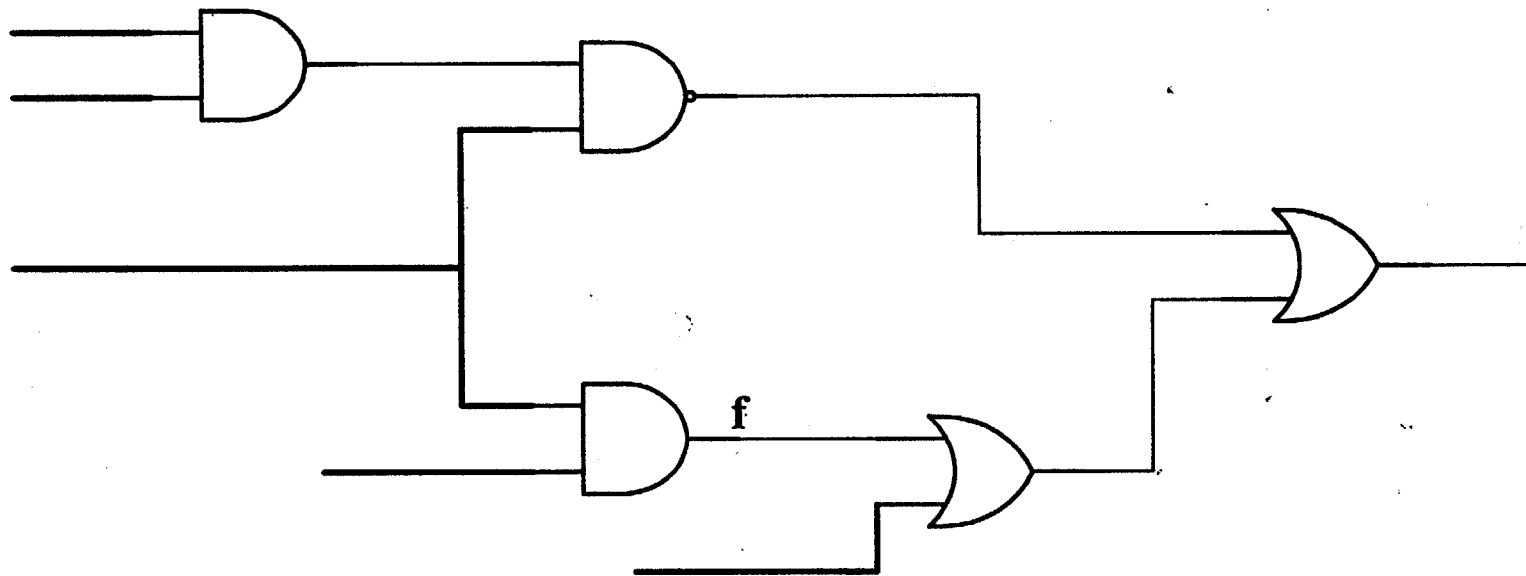
Fault Dominance Relation

Thm: In a fanout-free combinational circuit C , any test set that detects all SSF's on the primary inputs of C detects all SSFs in C .

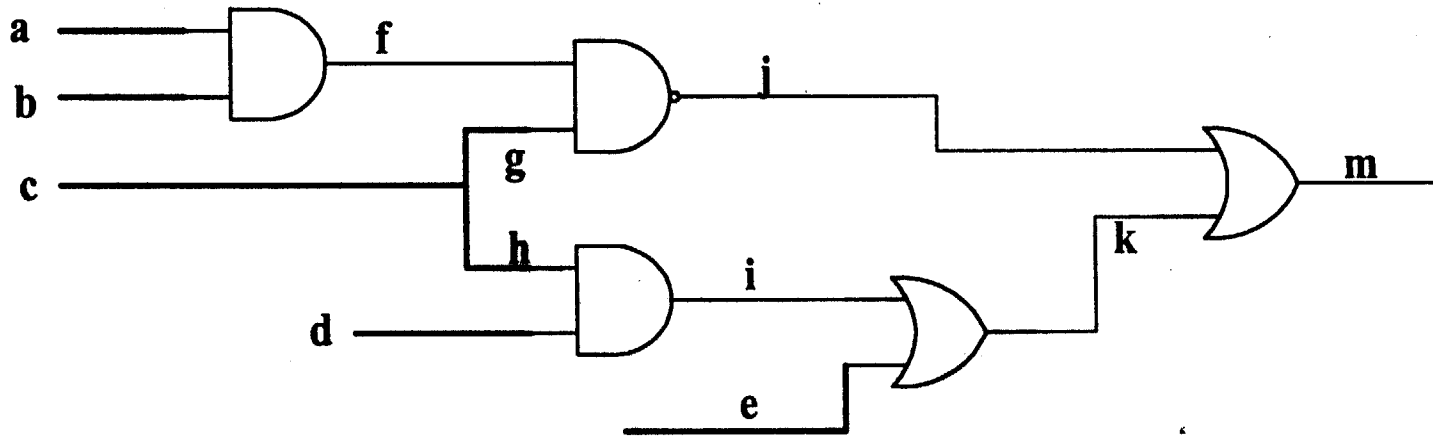
proof: use contradiction.

Thm: In a combinational circuit C, any test set that detects all SSF's on the primary inputs and the fanout branches of C detects all SSFs in C.

proof : use contradiction



Example :



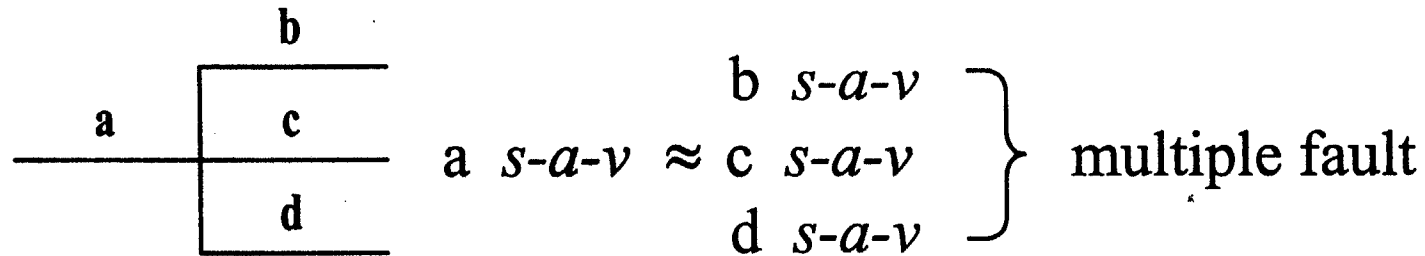
a	s-a-0	s-a-1	equivalent	dominate
b	s-a-0	s-a-1		
c	s-a-0	s-a-1		
d	s-a-0	s-a-1		
e	s-a-0	s-a-1 → i: s-a-1	equivalent	dominate
g	s-a-0	s-a-1 → f: s-a-1		
h	s-a-0	s-a-1		

- Total 12 lines, 24 faults
- 7 checking lines, 14 checking faults
- 24 faults have been reduced to 10 faults only

- Note :
- Checkpoint fault detection is only generated for irredundant circuits.
 - In redundant circuit, some of the checking faults are undetectable.
 - If we consider only checkpoint faults and we generate a test set that detects all detectable ^{Checkpoint} faults, this test set is not guaranteed to detect all detectable SSFs of the circuit; in such a case, additional tests may be needed to obtain a complete detection test set.

(∵ redundant faults)

The relations between faults on a stem line and on its branch lines



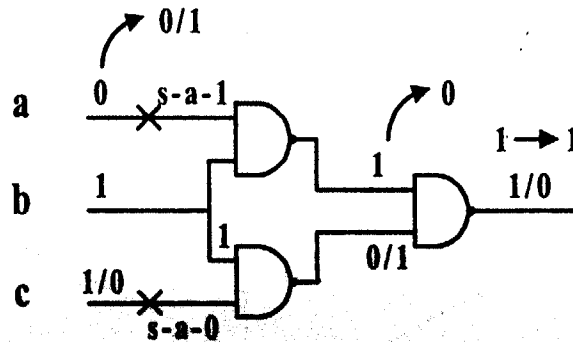
How about a and individual branch?

\Rightarrow No equivalence or dominance relations

Multiple Stack-at Fault model

- Masking relations among faults.

Def : Let T_g be the set of all tests that detect a fault g , we say fault f functionally masks the fault g if and only if $\{f, g\}$ is not detected by any test in T_g .

Example:

$a b c = 0 1 1$ is the only test pattern for $c s-a-0$.

$$\therefore T_{c s-a-0} = \{ 0 1 1 \}$$

Now, if a $s-a-1$

$\Rightarrow T_{c s-a-0}$ can't detect $c s-a-0$

\therefore a $s-a-1$ masks $c s-a-0$.

\therefore No single fault test set can guarantee to detect all multiple faults.

Note: If f masks $g \Rightarrow$ the fault $\{f, g\}$ is not detected by T_g

But $\{f, g\}$ may be detected by other tests.

Example: In the above example, $\{a\ s-a-1, c\ s-a-0\}$ can be detected by 010 which is not in $T_{c\ s-a-0}$.

So far: we know that

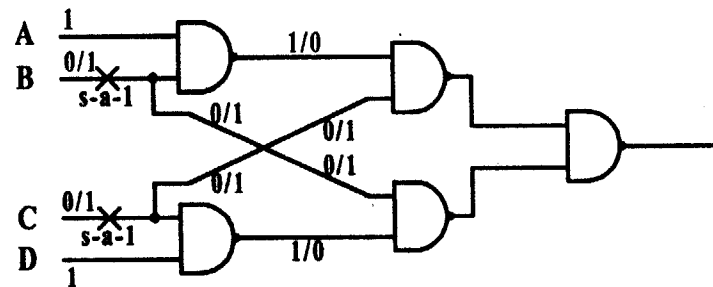
\exists multiple fault $\{f, g\} \Rightarrow \{f, g\}$ is not detected by T_g .

Question:

Given a complete single fault test set T , can there exist a multiple fault $F = \{f_1, f_2, \dots, f_k\}$ s.t. F is not detected by T ?

Answer: yes.

Circular Masking

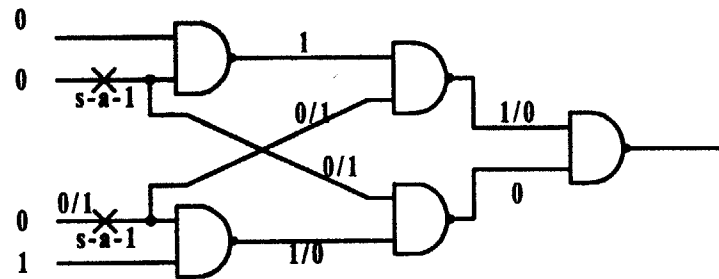
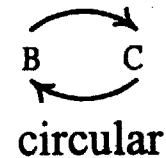


under test pattern $A B C D = \{ 1 0 0 1 \}$

B s-a-1 masks C s-a-1

C s-a-1 masks B s-a-1

B and c are circular masking



Circular masking can result in undetectable multiple stack-at faults (Multiple-line redundancy)

Example: Page 120 Breuer.

Note: circular masking is a necessary but not sufficient condition for undetection.